

Research Paper

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Circular polarization realization in array antennas by a modified sequential phase feeding mechanism of linearly polarized elements

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Abstract

Novel designs of a circularly polarized 2×2 and 4×4 slot array antennas (CPSAAs) are proposed. Sequential phase feed network composed of a section of 270° and four strips is utilized to feed the array structures. Array elements are simple linearly polarized wide slot antennas (WSAs) composed of simple radiating patches with truncation on four corners and slotted ground plane on substrate backside. Each WSA operates over the UWB frequency band of 3.1–10.6 GHz. When the WSAs are arranged in the form of 2×2 and 4×4 CPSAAs, frequency bands of 3–13.1 and 1.4–12.3 GHz are covered, respectively. More importantly, CP is generated at 4.5–8 and 4.6–8.8 GHz for 2×2 and 4×4 CPSAAs, respectively. This is while; the WSAs are linearly polarized elements. Higher gain values and wider bandwidths are obtained with respect to single WSA. Design process and performance analysis of the single and array antennas are discussed through the paper.

Introduction

In recent years, array antennas have attracted the attention of researchers due to their suitable performance in various possible communication applications. In this regard, paving the way to reach a highly reliable antenna system, many efforts have been devoted to design array antennas. Different applications such as beam switching [1], side lobe suppression [2], solid angle scanning [3], and millimeter-wave application [4] are envisaged for array antenna structures. On the other hand, circularly polarized (CP) antennas exhibit better performance comparing with linearly polarized (LP) ones in terms of better mobility and penetration features [5, 6]. As known, CP is generated when two orthogonal electric field vectors are excited on the antenna body. A vast variety of techniques could be applied to generate the aforementioned orthogonal vectors. For instance, the authors in [7] propose a high gain omnidirectional CP antenna for WLAN applications. Simple feed line is adopted to excite the antenna. To realize the CP, a helix-shaped slot is etched on outer conductor of the antenna. In [8], a low-profile CP antenna is proposed with wide AR bandwidth. Two orthogonal pairs of parallel slots etched symmetrically onto a ground plane are utilized as the antenna ground plane, which yield CP achievement. Apart from simple antennas with CP feature, the combination of reconfiguration technology with simple antennas results in interesting observations. This is due to the fact that by embedding switches on the antenna body, more than one operating cases could be attained which extends the antenna functionality considerably. An example of such a configuration is proposed in [9] within which small isosceles right angle triangular sections are removed from diagonally opposite corners for the generation of CP with axial ratio bandwidth of 11.1%. Moreover, by switching PIN diodes across the slits ON and OFF, reconfigurable CP antenna is realized. Dual-band behavior with CP characteristics is exhibited by the proposed antenna. Based on the above discussion, to benefit the merits of both CP and array antennas, CP array antennas (CPAAs) have been regarded as a popular category of antennas exhibiting remarkable characteristics. Although many CPAAs have been proposed in the literature, there are still opportunities to enhance the performance of such antennas [10–13]. For instance, utilizing simple array elements to be in line with techno-economic issues, proposing novel feeding strategies, and tuning the elements and their spacing such that the operating bands could be focused on applicable and mostly used frequency bands are some of the related topics.

Being stimulated by the aforementioned points, this paper aims at proposing novel 2×2 and 4×4 microstrip-fed slot array antennas with CP characteristics (CPSAAs). The single antennas utilized as the array elements are LP wide slot antennas (WSAs). These antennas are composed of a corner-truncated radiating patch on top and bottom edges with slotted

ground plane. Moreover, three semi-circular-shaped slots are removed from the ground plane to further enhance the single-antenna performance. A simple rectangular feed line is also adopted to feed the antenna structure. By suitably arranging four and 16 samples of the aforementioned WSAs, 2×2 and 4×4 CPSAAs are obtained. It is worth mentioning that the proposed constituent single antennas are LP elements. This is while; by arranging the WSAs in an array structures, CP is generated. In fact, CP is achieved by proper combination of LP elements and finely tuning their spacing and dimensions. Analyzing the WSA and CPSAAs performance reveals that the proposed CPSAAs exhibit better performance including higher gain, wider bandwidth, and CP property with respect to WSAs. Specifically speaking, the proposed design offers the following advantages:

- Achieving CP feature through suitable arrangement and proper combination of LP elements.
- Utilizing simple structured elements with cost-effective and compact design.
- Adoption of a novel feeding mechanism which enhances both 10 dB impedance bandwidth and 3 dB AR bandwidth.
- Suitable performance in the form of both 2×2 and 4×4 designs.

This paper is organized in the following sections: section “Design of array element” studies the design process of WSAs as the array elements. Section “Feeding method” surveys the topology and performance of the proposed feeding network used for the array structures. In section “ 2×2 CPSAA design and performance”, the performance of the proposed 2×2 CPSAA is studied in detail. Besides, simulated and measured results are studied and compared in depth. A comparison of the proposed 2×2 CPSAA configuration with some previous designs is carried out in section “Comparison”. Structural analysis and performance of 4×4 CPSAA is presented in section “ 4×4 CPSAA design and performance”. Ultimately, section “Conclusion” concludes the paper.

Design of array element

This section establishes the design process of the single-antenna element, which is utilized as the array element. The configuration of the proposed antenna is shown in Fig. 1. It is seen that the antenna is composed of a simple feed line with a corner-truncated radiating patch on top side of the substrate. On backside, a simple ground plane with a wide rectangular slot is embedded. Three semi-circular-shaped slots are removed from the lower edge of the ground plane to further enhance the antenna performance. FR4 substrate with a total size of $20 \text{ mm} \times 18 \text{ mm}$ and thickness of 1 mm is adopted as substrate. All antenna parameter values are reported in Fig. 1. To shed light on design process and performance enhancement process of the WSA, final antenna geometry achievement is discussed in a four-step process shown in Fig. 2. First, a simple rectangular feed line with a rectangular patch on top side and a slotted ground plane on backside is adopted (Ant. 1). In the sequel, three semi-circular slots are removed from the ground plane structure (Ant. 2). Then, two triangular-shaped slots are removed from the lower edge of the radiating patch (Ant. 3). Finally, top edge of the radiating patch is also truncated to achieve the final configuration (Ant. 4). The aforementioned antenna structures are simulated using Ansoft High Frequency Structure Simulator (HFSS) and the

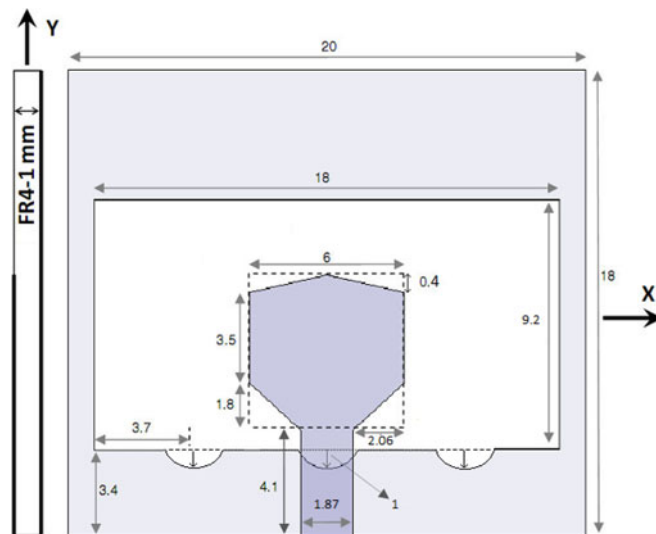


Fig. 1. Single WSA configuration and parameters.

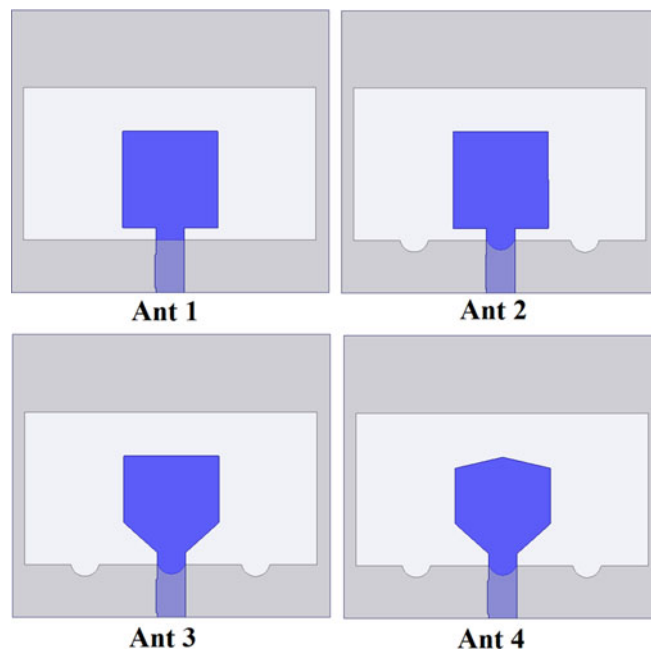


Fig. 2. Design steps of the single WSA.

corresponding S_{11} curves are plotted in Fig. 3. As the results clearly indicate, the operating frequency bands of Ant. 1 are in 3–5.3 and 8.6–10.1 GHz. Regarding Ant. 2, it is seen that by removing the semi-circular slots, the frequency band of 3.1–4.1 and 7.8–10.7 GHz is covered. The bandwidth is extended to higher frequencies. In the sequel, in Ant. 3, the truncation of the lower edge of the ground plane yields the impedance-matching improvement over the entire frequency band. This modification results in the coverage of 3.5–4.3 and 5.1–10.4 GHz. Finally, Ant. 4 operates over the frequency band of 3.1–10.6 GHz, which is exactly in line with UWB frequency band. As well, impedance matching is enhanced at 4–6 GHz.

It is worth noting that the embedment of slots does not excite new modes, but modifies the higher order orthogonal resonance

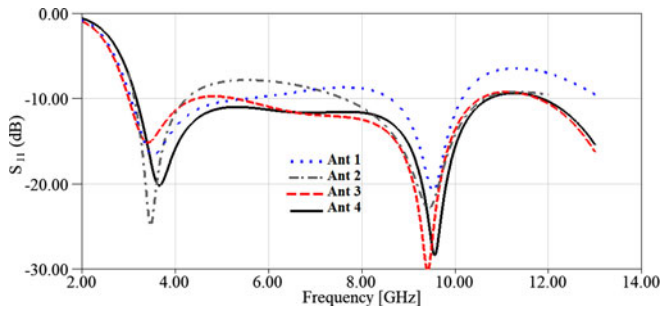


Fig. 3. S_{11} curves for the four design steps of single WSA.

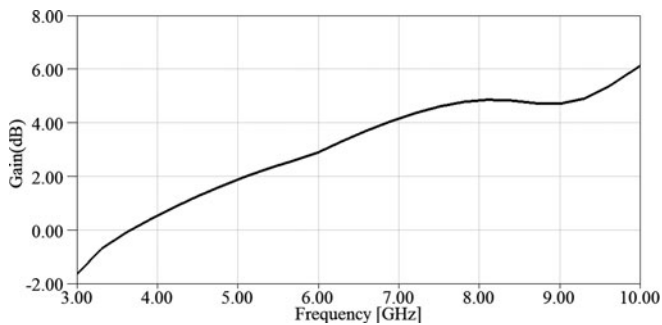


Fig. 4. Simulated gain of the proposed WSA.

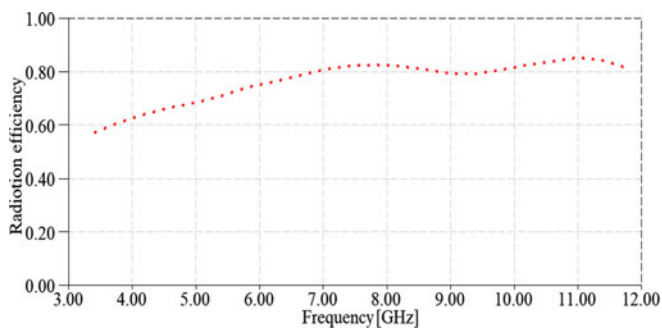


Fig. 5. Simulated radiation efficiency of the proposed WSA.

frequency in microstrip antennas. The resonance frequency is a function of slots dimensions and location, as follows [14]:

$$f = \frac{c}{2 L_e \sqrt{\epsilon_{re}}}, \tag{1}$$

where L_e is the electric length of the slot and ϵ_{re} relates to the antenna material. WSA gain is plotted in Fig. 4. It is seen that gain values increase up to 6 dB over the operating frequency band. As well, based on the obtained results in Fig. 5, an efficiency between 60 and 80% is observed over the operating frequency band. Moreover, radiation pattern at the resonance frequency of 9.5 GHz is illustrated in Fig. 6. As can be seen, the obtained radiation patterns are in line with communication systems requirements.

To shed light on the advantages of the proposed single antenna with respect to some of the previously designed antennas, Table 1 compares the structural and performance of the presented

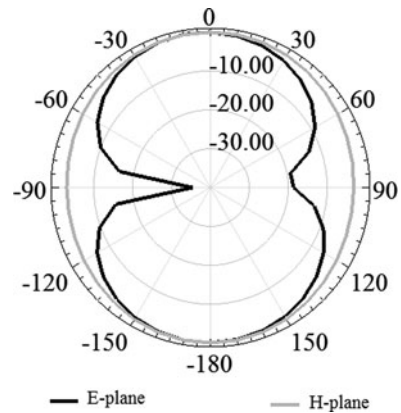


Fig. 6. Radiation pattern of WSA at 9.5 GHz.

WSA and some other antennas. The comparison is based on the antenna total size, 10 dB impedance bandwidth, and peak gain values. As can be seen, the single WSA is the smallest one among the antennas. This is while the candidate antennas performance comparison reveals that the proposed design is ranked between the antennas with widest bandwidth. Although having smaller size which yields less freedom degree, the proposed WSA provides a high and suitable gain value.

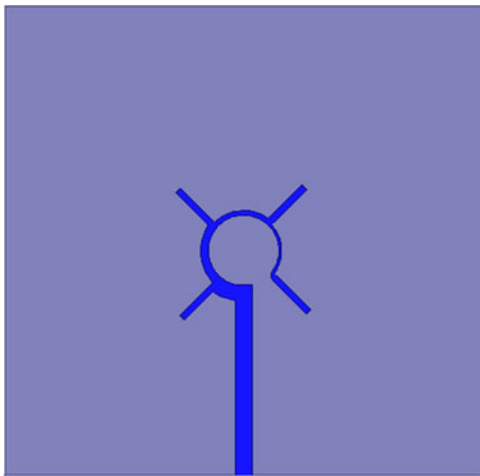
Feeding method

Arranging the aforementioned WSAs in array configuration calls the need for an efficient feeding system, which could suitably meet the requirements of the array antenna. In single antennas, usually a simple feed network provides the needed features of the antenna system, but this issue is not a simple matter in array antenna structures. As an array antenna consists of hundreds of thousands of single antennas, feeding strategy selection becomes a more important factor, which needs careful and meticulous considerations. Designing feed networks has been the main focus of many studies presented so far. One of the most interesting and capable ones is the sequential phase (SP) feed network. Different forms of SP feed and corresponding results have been surveyed, which although end in applicable and interesting results, still are faced with some challenges. Accordingly, more trustworthy solutions are required. It has been shown that adoption of SP feed in 2×2 CPAAs enhances the array functionality to a great extent. Detailed discussion on design of SP feed could be accessed in [22]. With the aim of benefiting the advantages of SP feed network, a modified SP feed is deployed in the present design shown in Fig. 7. As can be seen, a sector of 270° with different radii in each comprising part and four strips are the constituting elements of the proposed SP feed network. Due to the inherent features of this feed mechanism, different phases with stable difference are excited at the end of each strip, which yields CP generation. As mentioned earlier, SP feed network facilitates polarization purity increment, impedance matching, and pattern symmetry in CPAAs.

To scrutinize the advantages of the proposed feeding technique over the simple SP feed network, two antennas with simple and proposed feeding configurations are studied. The simple SP feeding method consists of a sector of 270° with the same widths in constituent sections, named as Ant. 1 in Fig. 8. As can be seen, Ant. 2 is the antenna with the proposed feed network in this

Table 1. Comparison of the proposed WSA with some previously designed antennas

Ref.	Size	10 dB impedance bandwidth	Peak gain
[15]	40 × 40 × 1.52	2.57–4.16 (47%)	3.56 dBic
[16]	170 × 150 × 1.6	1.102–2.168 (65%)	5.8 dBic
[17]	80 × 80 × 0.8	2.4–6.22 (89%)	6 dBic
[18]	60 × 40 × 1.6	2.1–7.5 (113%)	3.8 dBi
[19]	75 × 75 × 0.8	2.15–2.62 (20%)	8 dBi
[20]	60 × 60 × 1.6	2–5 (86%)	4.5 dBic
[21]	60 × 60 × 1.6	1.6–5.25 (107%)	4 dBic
This work	20 × 18 × 1	3.1–10.6 (110%)	6 dBi

**Fig. 7.** SP feed network configuration of the proposed 2 × 2 CPSAA.

paper. The simulation results are shown in Figs 9 and 10 for S_{11} and AR bandwidths, respectively. The results clearly indicate that both S_{11} and AR bandwidth are enhanced significantly by the application of proposed feeding mechanism, that is, the utilization of different widths in different sections of the sector of 270°. It is seen that the adoption uniform widths yield no CP, this is while; by modifying the widths of the sections, wide CP is obtained in 4.5–8 GHz frequency band.

2 × 2 CPSAA design and performance

Design process and performance of the WSA was discussed in the previous section. To further enhance the antenna performance, the proposed element is arranged in the form of a 2 × 2 CPSAA excited with the aforementioned SP feed network. As clearly shown in Fig. 11, four WSAs are connected to the four strips originated from the SP feed line. On top side of the substrate, the SP feed network and four corner-truncated patches are printed. Besides, on backside, four wide slots are included. The overall size of the 2 × 2 CPSAA is 50 mm × 50 mm × 1 mm which is more compact with respect to many of the previous designs. Moreover, detailed values of the proposed SP feed are reported in Table 2.

To analyze the proposed 2 × 2 CPSAA performance in real applications, an experimental prototype is fabricated and measured. Top and bottom views of the fabricated antenna are

shown in Fig. 12. As it was mentioned in WSA design, FR4 material with a thickness of 1 mm is adopted as the substrate. Moreover, the fabricated prototype under measurement process is shown in Fig. 13. Simulated and measured S_{11} curves for the proposed 2 × 2 CPSAA are illustrated in Fig. 14. Close investigation of S_{11} curve performance reveals considerable enhancement with respect to the WSA. As it was shown in Fig. 3, the single WSA operates over the frequency band of 3.1–10.6 GHz. This is while; the 2 × 2 CPSAA covers the frequency band of 3–13.1 GHz. Moreover, agreement of the simulated and measured results confirms the CPSAA suitable performance. As an important point, it should be noticed that the single array elements, namely WSA, are LP elements, but when arranged in the form of 2 × 2 SAA, exhibit CP characteristics. Simulated and measured AR and gain values shown in Fig. 15 are highly compatible with each other, which confirm the outperformance of the proposed design. As clearly seen, CP is obtained in 4.5–8 GHz within which gain values vary between 6 and 8 dBic. Simulated radiation efficiency of the proposed array antenna is plotted in Fig. 16. As can be seen, the efficiency varies between 80 and 90% over the obtained frequency band of 4.5–8 GHz where CP is observed. Moreover, right-hand circular polarization (RHCP) and left hand circular polarization (LHCP) radiation patterns at 4.2 and 8.3 GHz are plotted in Fig. 17. Moreover, about 20 dB difference is observed between the Co and cross polarization values.

With the aim of more discussion on the proposed 2 × 2 CPSAA antenna performance, surface current distribution at the frequency of 4.5 is shown in Fig. 18. The results are depicted at 0, 90, 180, and 270° phases. The direction of the dominant current at each phase is shown by red arrows. Following the direction of currents confirms RHCP at +z direction and LHCP at -z direction.

Comparison

Comparison of the proposed performance of the 2 × 2 array structure with some of the previous designs would be fruitful to shed light on the advantages of the present work. To this end, Table 3 provides the characteristics of some previously published designs in [23–28] and the proposed antenna. As the provided data imply, the proposed antenna is the smallest among the antennas in [23–27]. As can be seen, the antenna in [28] is smaller. This is while; the widest 10 dB impedance bandwidth and 3 dB AR bandwidth are obtained by the present work. Regarding the peak gain values, it is seen that the proposed design in this work offers an acceptable and high gain. As well, bidirectional radiation pattern is

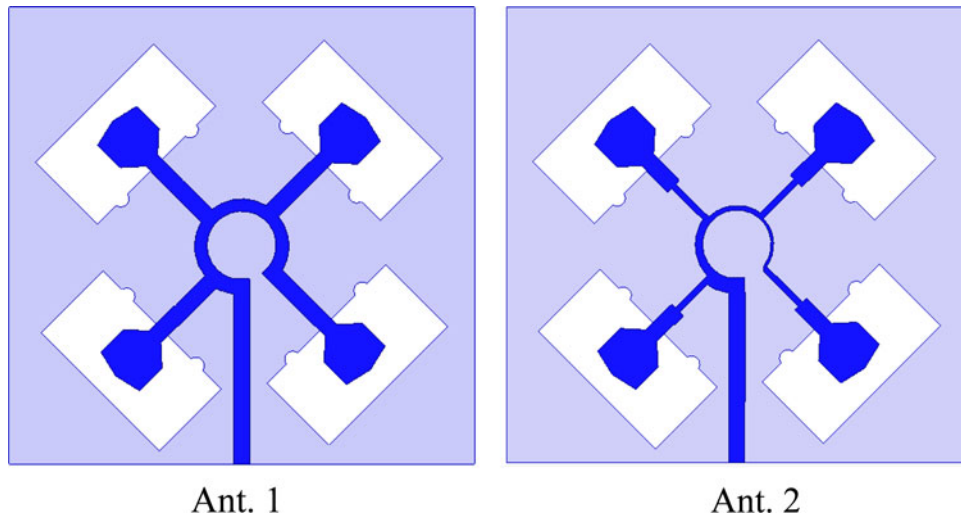


Fig. 8. Two configurations of feed networks.

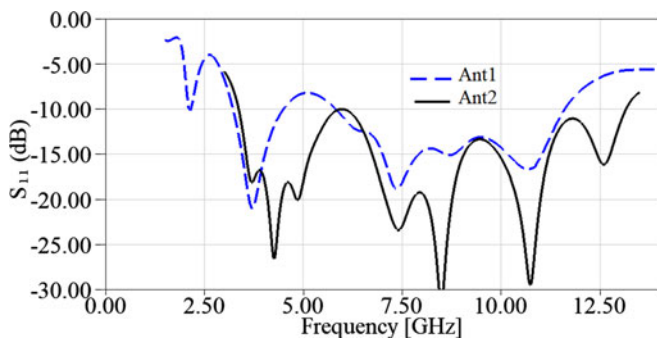


Fig. 9. S_{11} curves for Ant. 1 and Ant. 2 with different feeding configurations.

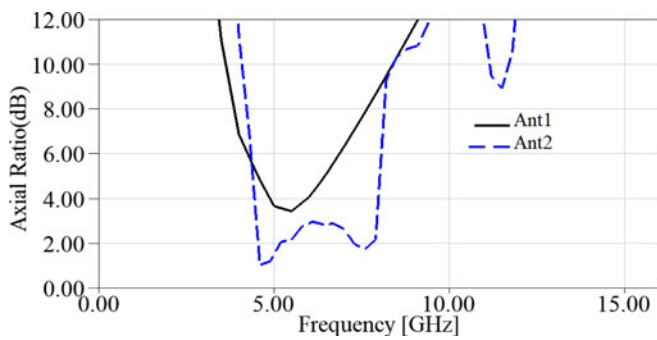


Fig. 10. AR curves for Ant. 1 and Ant. 2 with different feeding configurations.

observed for the antenna. It is clearly seen that compact size, and wide impedance and AR operating bandwidths are of the evident merits exhibited by the array antenna in this paper.

4 × 4 CPSAA design and performance

Further performance enhancement could be realized by increasing the number of single WSAs in a 4 × 4 array. Sixteen single WSAs are suitably arranged in the form of an array antenna with enhanced performance with respect to both WSA and 2 ×

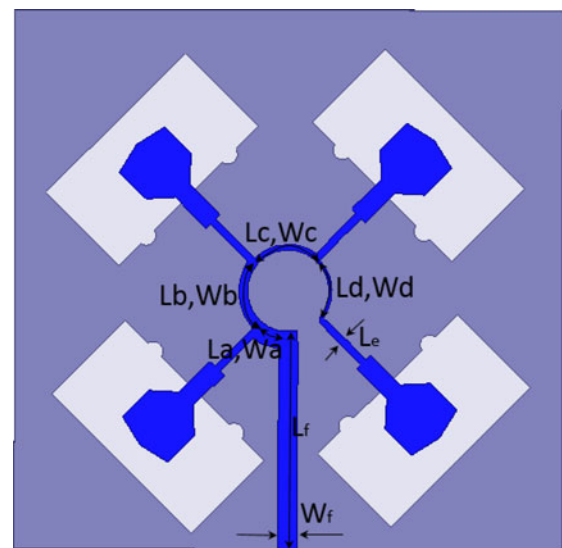


Fig. 11. SP feed network configuration of the proposed 2 × 2 CPSAA.

Table 2. SP feed network parameters values (all parameters are in mm)

L_f	W_f	L_a	W_a	L_b	W_b
20.35	1.85	2.65	1.8	6.5	0.9
L_c	W_c	L_d	W_d	L_e	
6.5	0.45	6.5	0.22	0.7	

2 CPSAA. Figure 19 illustrates the geometry of the proposed 4 × 4 CPSAA. As seen, SP feed network, as discussed in previous sections, is adopted as the feeding strategy. The fabricated prototype of the 4 × 4 array and the antenna under measurement are illustrated in Figs 20 and 21, respectively.

Simulated and measured S_{11} curves for this array structure are shown in Fig. 22. It is seen that the performance is remarkably enhanced specially at lower frequencies. The agreement of the

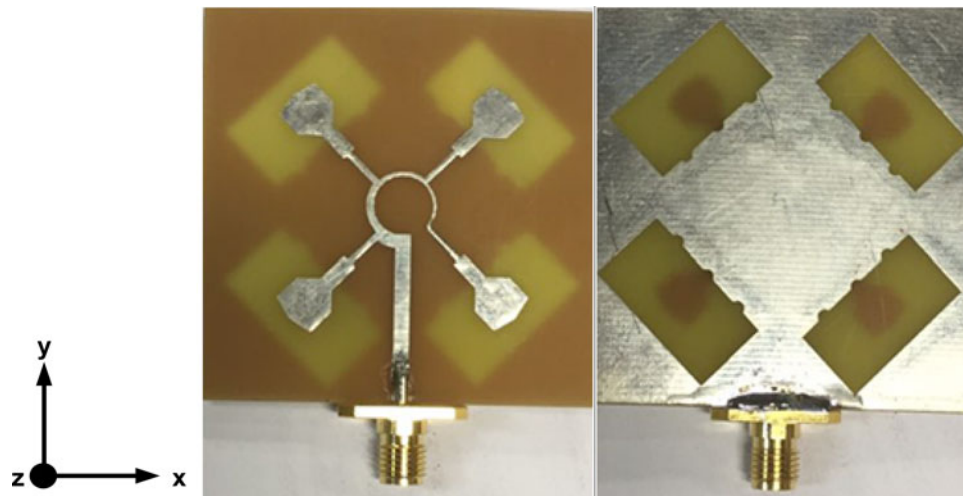


Fig. 12. Top and bottom views of the fabricated prototype.

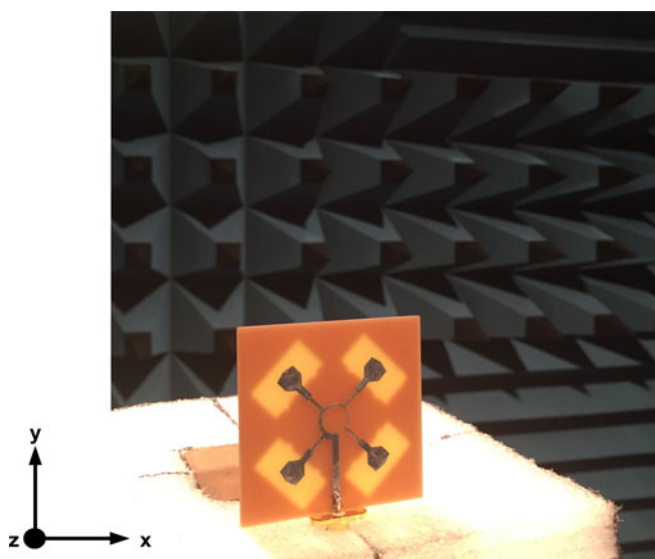


Fig. 13. The fabricated prototype of 2×2 CPSAA in measurement process.

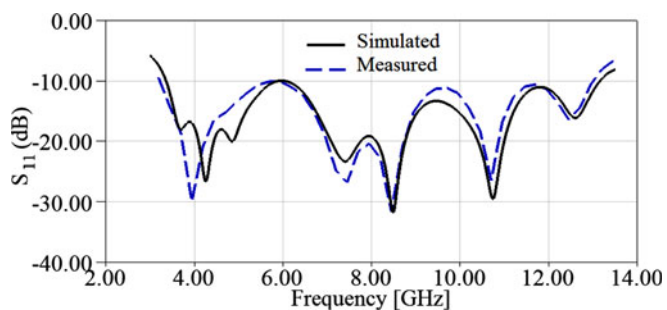


Fig. 14. Simulated and measured S_{11} curves for the proposed CPSAA.

simulated and measured results confirms that the frequency band of 1.4–12.3 GHz is covered by the array. As well, simulated and measured AR curves are compared in Fig. 23. The results reveal CP generation at 4.6–8.8 GHz, which is better than the results

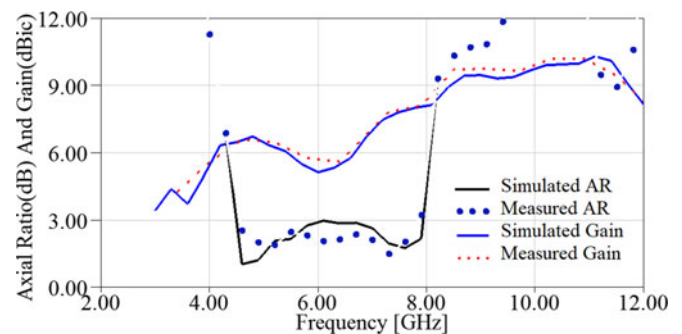


Fig. 15. Simulated and measured AR and gain curves for the proposed CPSAA.

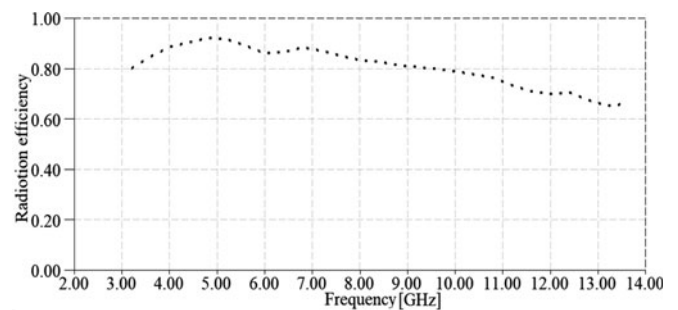


Fig. 16. Simulated radiation efficiency for the proposed CPSAA.

obtained for 2×2 CPSAA. Moreover, simulated and measured gain results are depicted in Fig. 24. A maximum gain of 11 dB is observed for the proposed array over the operating frequency band.

Radiation pattern of the proposed 4×4 CPSAA is depicted in Fig. 25 at 5 and 7.5 GHz for $\phi = 0$ and 90° . It is observed that a suitable radiation pattern for communication systems is achieved, which confirms the antenna outperformance in relevant applications.

To have a more precise record on performance improvements, the operational characteristics of the proposed single WSA, 2×2 CPSAA, and 4×4 CPSAA are summarized in Table 4. Based on

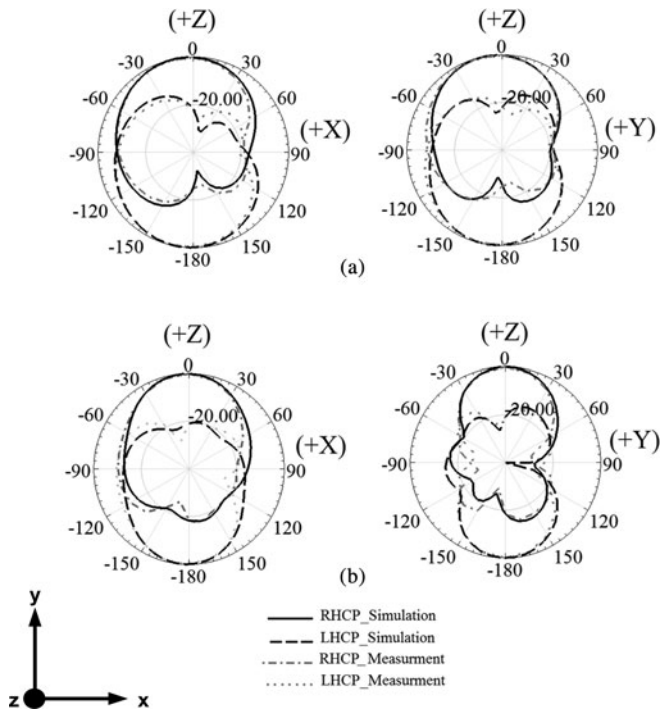


Fig. 17. Simulated and measured RHCP and LHCP radiation patterns of the proposed 2 × 2 CPSAA, (a) 4.2 GHz and (b) 8.3 GHz.

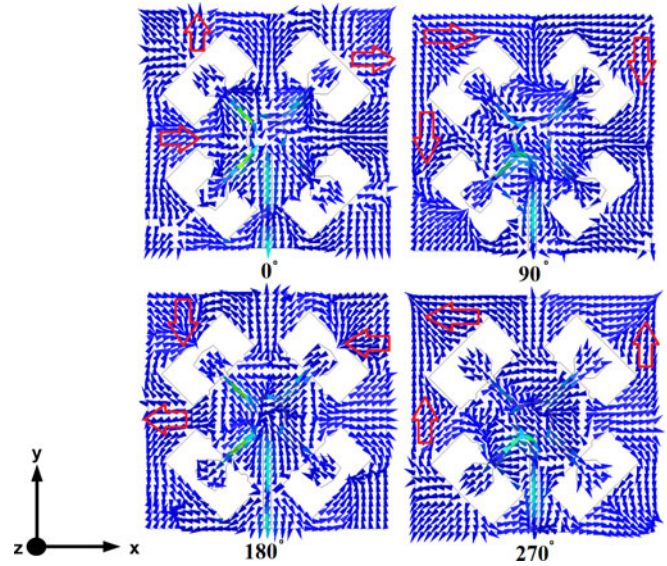


Fig. 18. Surface current distribution on the antenna at 4.5 GHz.

Table 3. Summary of the characteristics of some previously designed structures and the present work

Ref.	Size (mm ³)	10 dB impedance bandwidth	3 dB AR bandwidth	Peak gain	Radiation pattern status
[23]	90 × 90 × 0.8	4–8.1	4.75–6.82	8 dBi	Bidirectional
[24]	100 × 100 × 3	4.82–5.12	4.84–5.13	10.5 dBic	Unidirectional
[25]	75 × 75 × 1.5	5.2–6.23	5.25–5.95	12 dBic	Unidirectional
[26]	220 × 220 × 1.6	1.15–1.95	1.15–1.9	8 dBi	Bidirectional
[27]	90 × 90 × 0.5	4.46–5.5	4.52–5.4	11.5 dBi	Unidirectional
[28]	45 × 45 × 1.6	5–6.75	5.32–6.22	8.2 dBi	Unidirectional
This work	50 × 50 × 1	3–13.1	4.5–8	10.2 dBic	Bidirectional

the data provided in Table 4, both 10 dB impedance bandwidth and 3 dB AR bandwidth are improved by increasing the number of used WSAs in array structures.

Conclusion

A simple design of simple single slot antenna with LP feature over the frequency band of 3.1–10.6 GHz was designed and discussed in this paper. The WSA configuration was composed of a simple radiation patch with truncation on the four corners. Besides, a slotted ground plane was included. Then, with the aim of performance enhancement, the aforementioned WSA was arranged in the form of 2 × 2 and 4 × 4 CPSAAs. A novel design of SP feed network, which is composed of a sector of 270° with unequal widths in three sections along with four strips, feeds the array structure. The proposed CPSAAs operate over the frequency bands of 3–13.1 GHz with CP at 4.5–8 GHz and 1.4–12.3 GHz with CP at 4.6–8.8 GHz, respectively, based on 2 × 2 and 4 × 4

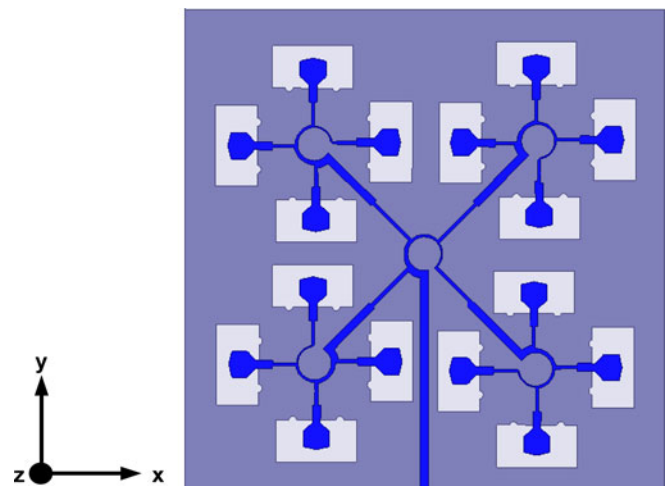


Fig. 19. Geometry of the proposed 4 × 4 CPSAA.

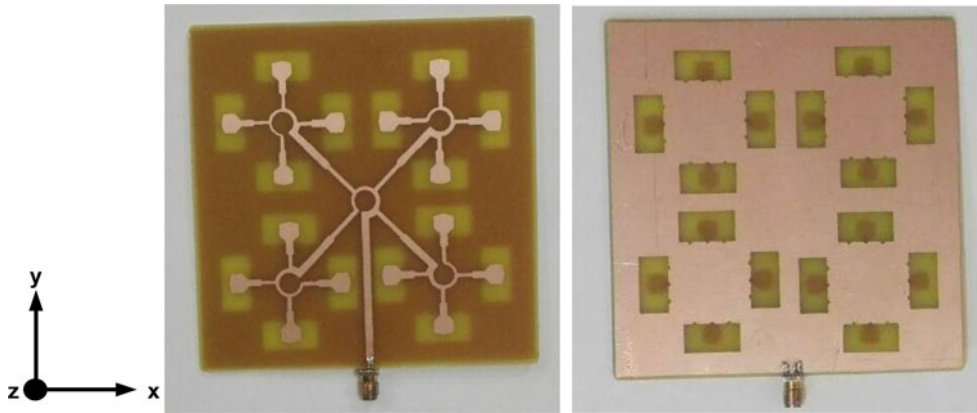


Fig. 20. Top and bottom view of the fabricated prototype of the 4×4 CPSAA.

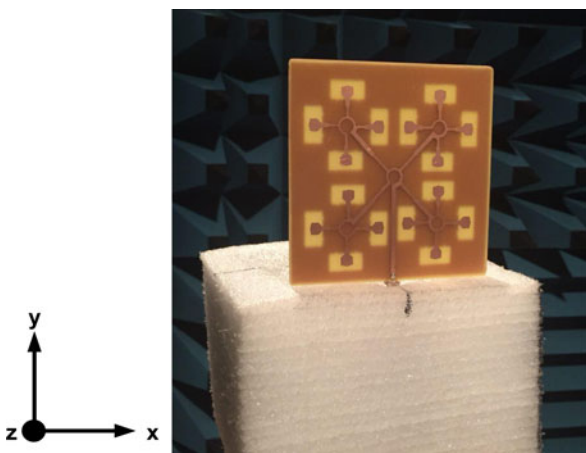


Fig. 21. 4×4 CPSAA under measurement process.

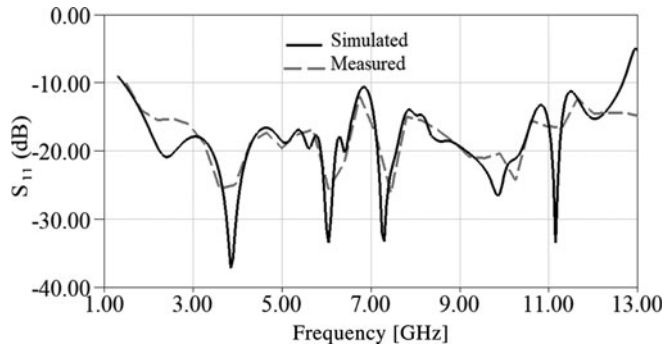


Fig. 22. Simulated and measured S_{11} curve for the proposed 4×4 CPSAA.

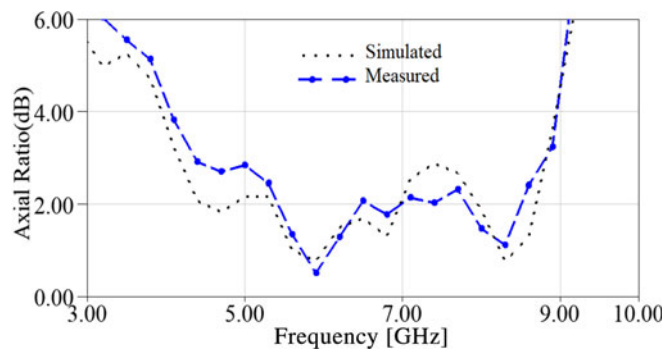


Fig. 23. Simulated and measured AR curve for the proposed 4×4 CPSAA.

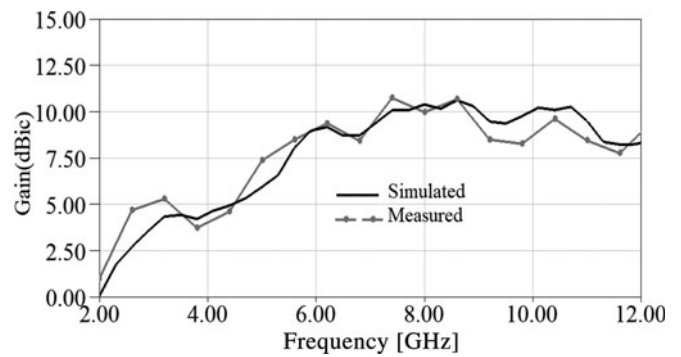


Fig. 24. Simulated and measured gain curve for the proposed 4×4 CPSAA.

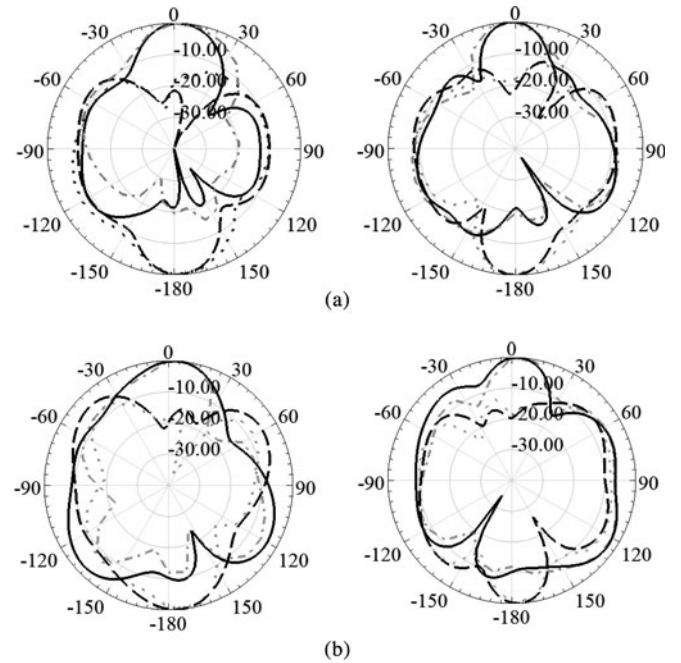


Fig. 25. Simulated and measured radiation patterns of the proposed 4×4 CPSAA, (a) 5 GHz, (b) 7.5 GHz.

ones. Wide impedance and AR bandwidth, higher gain values, and good radiation properties are some of the merits of the proposed CPSAAs.

Table 4. Structural and performance characteristics of the proposed arrays and array element

Proposed antenna	Size (mm ³)	10 dB impedance bandwidth (GHz)	3 dB circular polarization (GHz)
WSA	20 × 18 × 1	3.1–10.6	No CP
2 × 2 CPSA	50 × 50 × 1	3–13.1	4.5–8
4 × 4 CPSA	105 × 105 × 1	1.4–12.3	4.6–8.8

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