

RESEARCH PAPER

Fork-shaped planar antenna for Bluetooth, WLAN, and WiMAX applications

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A microstrip transmission line fed fork-shaped planar antenna is proposed for Bluetooth, WLAN, and WiMAX applications. The antenna made of a microstrip feed line, fork-shape patch on one side and defected ground plane on the other side of dielectric substrate. A fork-shape is formed by two side circular arms and a rectangular central arm. The inverted T-shaped ground plane with a rectangular slot in the center arm is used to increase the bandwidth with better impedance matching of the lower band. The antenna is practically fabricated to validate the design. The antenna resonates dual band to cover an entire the WLAN and WiMAX bands. The antenna shows the measured bandwidth of 410 MHz (2.26–2.67) and 3.78 GHz (3.0–6.78 GHz) at lower and upper bands, respectively.

Keywords: Fork-shaped, Dual-band antenna, Slot antenna, WiMAX/WLAN applications

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I. INTRODUCTION

In the last decades, there has been remarkable development in wireless communication industry because it provides much more flexible way of communication. It is advantageous for a single wireless system to have an access to several services in which two or more bands with acceptable separation are required. Normally a dual-band or multiband antenna is required to fit in many services in one device such as wireless local area network (WLAN) and worldwide interoperability for microwave access (WiMAX) as these two techniques are now widely used in wireless communication devices. The operating bands for these technologies as assigned by IEEE 802.11 are 2.4 GHz (2.4–2.484 GHz), 5.2/5.8 GHz (5.15–5.35 GHz/5.725–5.825 GHz), and 2.5/3.5/5.5 GHz (2500–2690/3400–3690/5250–5850 MHz). In the previous years, many multiband antennas have been designed and reported such as H-shaped slot antenna [1], asymmetric M-shaped antenna [2], diversity monopole antenna [3], and ear-shaped antenna [4]. A defected ground structure antenna was presented for dual-band application [5] and an antenna consists of a circular ring, a Y-shape-like strip, and a defected ground plane is used to excite triple band [6]. An arrowhead-shaped slotted microstrip antenna is designed to achieve circular polarization [7]. An antenna comprises of a rectangular center strip, two lateral strips, and excited by microstrip line can be used to generate two separate bands to cover WLAN [8]. The antenna consists of planar loop element with a shorting pin is proposed for 2.45/5-GHz WLAN and 2.1-GHz wideband code-division multiple access (WCDMA)

applications [9]. Further, planar inverted-F antenna (PIFA) in conjunction with a parasitic element is presented for WLAN operations [10], an inverted L-slot triple-band antenna is presented in [11], and a CPW-fed tri-band printed antenna is proposed in [12].

In this paper, a novel dual-band fork-shaped radiator patch with a defected ground plane is proposed for Bluetooth, WLAN, and WiMAX applications. The antenna is made up of a microstrip feed line, fork-shaped patch on one side and a defected ground plane on the other side of the dielectric substrate. A fork-shaped patch is formed by two side circular arms and a rectangular central arm. The ground is of an inverted T-shaped with a rectangular slot in the middle arm. Detailed geometry configuration and parametric studies of the antenna are demonstrated in this paper. The effects of the patch and ground dimensions on the antenna performance are also studied and presented. Table 1 illustrates the useful fact about the antennas recently proposed for WiMAX/WLAN applications. From Table 1, it is evident that the proposed antenna has smallest size and also covers Bluetooth band as compared with the other mentioned antennas with a sufficient bandwidth at both the bands to cover entire Bluetooth, WLAN, and WiMAX bands. In Section II, we describe the antenna design. In Section III, parametric studies of antenna parameters are presented. Finally, the conclusion is provided in Section IV.

II. ANTENNA STRUCTURE AND DESIGN

Figure 1 shows the geometry of the antenna, fabricated on a 1.6 mm thick FR4 substrate with permittivity and loss tangent of 4.4 and 0.02, respectively. The proposed antenna is composed of a planar I-shaped monopole and two branch

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Table 1. Performance comparison of the proposed antenna with other reported antennas.

Reference	Size (mm × mm)	Operating bands (GHz)
[1]	60 × 60	1.55–1.57, 2.39–2.69, 4.97–5.93
[2]	64 × 62	2.38–2.49, 3.49–3.63, 5.57–6.20
[3]	40 × 90	2.4–4.2
[4]	18 × 37	2.38–2.78, 3.28–3.76, 4.96–5.96
[5]	31.6 × 31.6	2.30–2.48, 3.80–3.9
[6]	38 × 25	2.4–2.7, 3.1–4.15, 4.93–5.89
[7]	87 × 87	0.888–0.923
[8]	10 × 50	2.19–2.52, 4.84–6.07
[9]	58 × 62	1.92–2.52, 5.0–6.2
[10]	16.5 × 11.5	2.36–2.52, 5.1–5.9
[11]	20 × 30	2.39–2.51, 3.15–3.91, 5–5.53, 4.91–6.08
[12]	23 × 36.5	2.33–2.76, 3.05–3.88, 5.57–5.88
Proposed	24 × 35	2.26–2.67, 3.0–6.78

lines of R_1 and R_2 the outer and inner radii, respectively, and then the upper edge of the annular ring is etched by a circle of radius R_3 . The two branch lines act as two monopole radiators at the first band and at the same time acts as a cavity resonator fed by the I-shaped monopole that radiates in the upper frequency bands. The two branch lines increase the path over which the surface current flows and that eventually results in lowering the resonant frequency. The electrical length of the two branches is 32.11 mm which is quarter wavelength at the desired narrow band resonance 2.26 GHz. This structure is fed by a 50 Ω microstrip transmission line of width (W_f) 2.2 mm and length (L_f) 12.5 mm on one side of the substrate. On the other side, inverted T-shaped ground plane is printed with a rectangular slot in center T-arm

Figure 2 illustrates the steps involved in the proposed antenna design by using the electromagnetic simulation tool ANSYS “HFSS” [13]. For the case of antenna 1 of circular shape (curve (i)), a single wide band appears over the

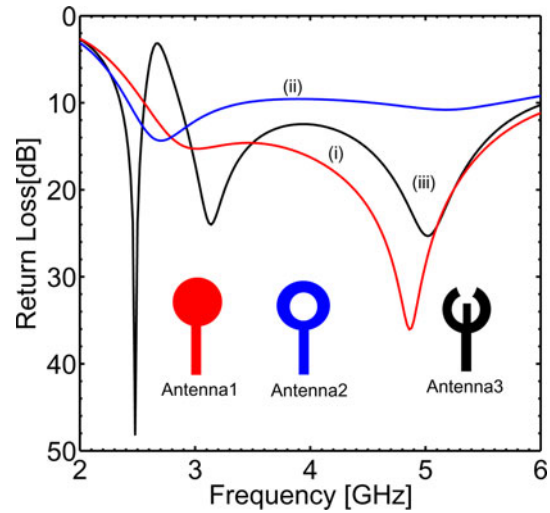


Fig. 2. Simulated return loss against frequency for circular, annular ring, and the proposed fork-shaped antenna.

frequency band, while the resonant mode seems to form at about 4.90 GHz. As for the case of antenna 2 of annular ring shape (curve (ii)), the band is still single. However, the mode is resonating at very low frequency, about 2.8 GHz. Finally, antenna 3 with inclusion of a circular slot is cut at the upper edge and the rectangular strip (curve (iii)) in the design will generate two separate bands first band from 2.24 to 2.68 GHz and second band from 3 to 6 GHz. It provides good performance and covers all the 2.4/5.2/5.8-GHz WLAN and 2.5/3.5/5.5-GHz WiMAX bands. Note that in these cases all the unmentioned dimensions are the same as listed in Table 2.

Based on the optimized parameters given in Table 2, a prototype for the proposed antenna was fabricated and

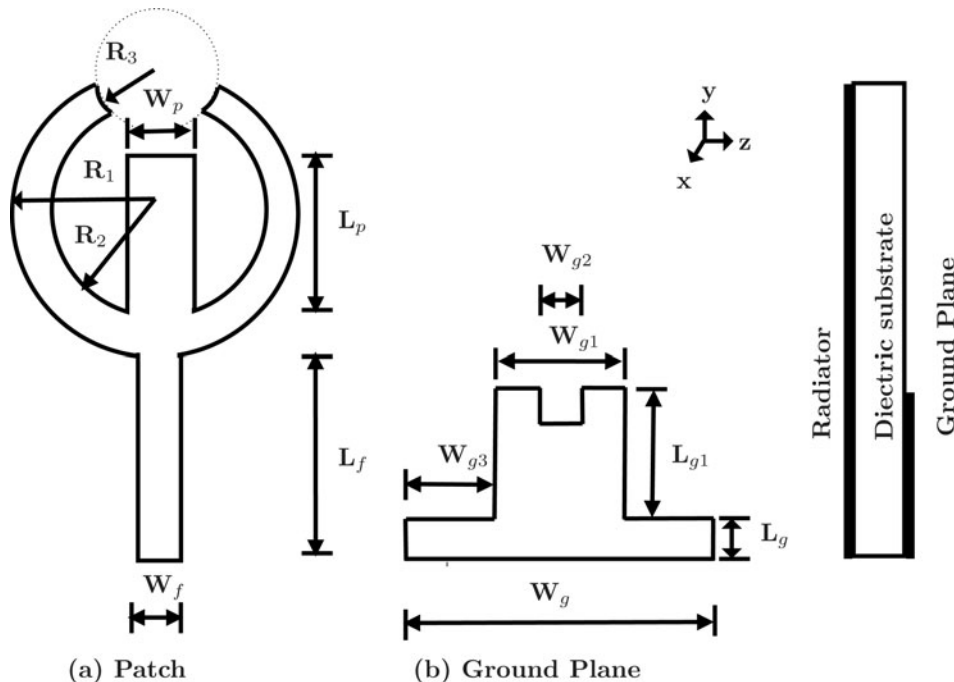


Fig. 1. Schematic configuration of the proposed fork-shaped antenna.

Table 2. Design parameters of the proposed antenna shown in Fig. 1.

Parameters	Unit (mm)	Parameters	Unit (mm)
L_p	13.9	L_g	2
W_p	4.4	L_{g1}	7
L_f	12.5	W_g	18
R_1	11.5	W_{g1}	13
R_2	8.9	W_{g2}	4
R_3	4.0	W_{g3}	2.5

tested. The photograph of the fabricated antenna is shown in Fig. 3. The simulated surface current distribution at different frequencies, 2.5, 3.5, and 5.5 GHz are shown in Figs 3(c)–3(e). It is observed from Fig. 3(c) that the strong surface current flows along the entire patch radiator and feed line. It is observed that the circular arm of fork-shaped radiator generates the lower resonant mode of 2.5 GHz. For the resonance at 3.5 GHz as shown in Fig. 3(d), it is found that the surface current is distributed over the central arm of the fork-shaped radiator and the middle portion of the ground plane. Furthermore, the results in Fig. 3(e) divulge that surface current of the third resonant mode (5.5 GHz) is mainly distributed on the feed line; the bottom part of the circular arm of the fork-shaped radiator; and the entire ground plane. Thus, it can be concluded that mainly the ground plane is accountable for this mode of resonance.

III. PARAMETRIC STUDY

All the major design parameters are used to find respective influence on the antenna performance. The outer radius R_1 , inner radius R_2 , radius of the defect to form a fork-shape R_3 , length L_{g1} , L_g , and width W_g of the ground plane are chosen for the parametric study. At a time, the single parameter is used to parametric study, while for other parameters the optimized values, as listed in Table 2, are used.

A) Patch parameters

Figure 4 shows the effect of radius R_1 of circular arm of the patch on the return loss of the antenna from 9.5 to 13.5 mm. It is found from Fig. 4 that it mainly affects the first band and the bandwidth of the second band as varied from 9.5 to 13.5 mm. It is also observed that the starting frequency of the second band is moving down as the radius varies up to 11.5 mm and again shifts toward the higher side afterwards. Thus, the optimum value of the radius R_1 of

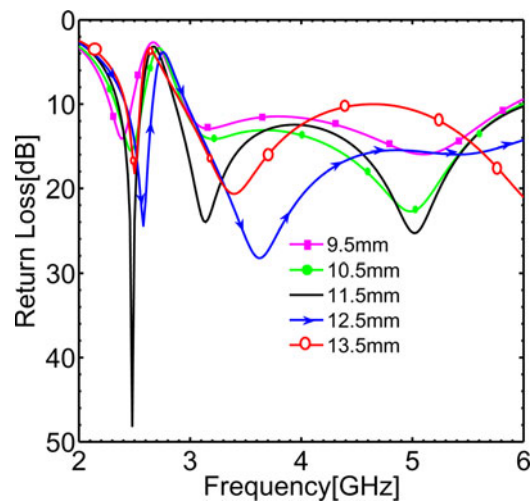


Fig. 4. Simulated return loss against frequency for the microstrip-fed fork-shaped antenna with various outer radius R_1 ; other parameters are the same as listed in Table 1.

the circular arm of the patch is chosen as 11.5 mm. The effect of inner radius R_2 of the circular arm of the patch on the performance of antenna is depicted in Fig. 5. It is observed that it decides the frequency of resonant mode of the first band. It can be seen that the resonant mode of the first band is shifted toward the lower frequency as varied from 6.9 to 10.9 mm. Thus, the optimum value for the inner radius R_2 of the fork-shaped patch is taken as 8.9 mm to get resonant mode of first band at 2.5 GHz. Figure 6 shows the influence of the radius of the circular defect R_3 over the antenna performance. It is observed that it mainly influences the frequency of the resonant mode of the first band and the starting frequency of the second band as varies from 2 to 6 mm. Therefore, the optimized value for the radius R_3 of the circular defect is chosen as 4 mm.

B) Ground parameters

Figure 7 shows the effect of length L_{g1} of the ground plane on the performance of the antenna. It is depicted that as the length L_{g1} is varied from 5 to 9 mm, it mainly influenced the frequency of the resonant mode of the first band and the starting frequency of the second band. Thus, optimum value of L_{g1} is chosen as 7 mm. The effect of width W_g of the ground on the performance of antenna is shown in Fig. 8. It is keenly observed that it also influenced the frequency of the resonant mode of the first band and the starting

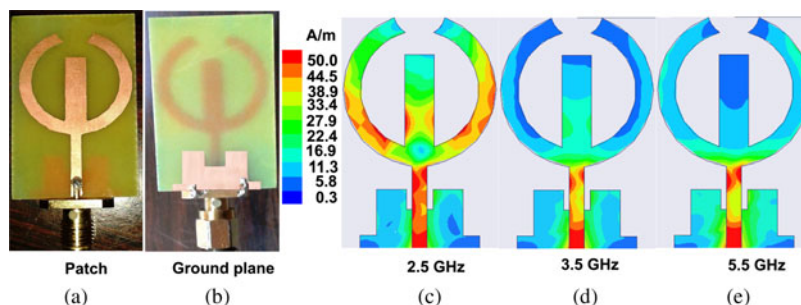


Fig. 3. Fabricated photograph and simulated surface current densities at various sampling frequencies of the proposed fork-shaped antenna.

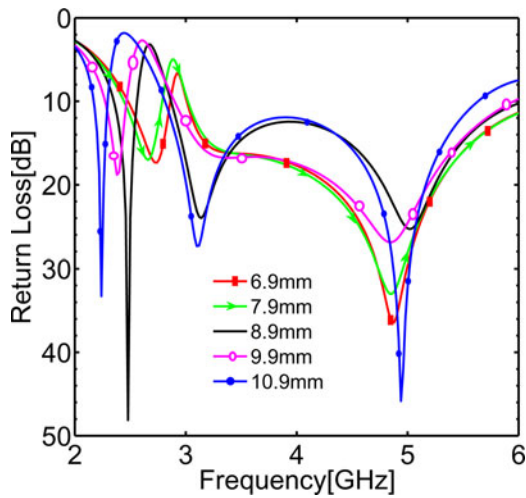


Fig. 5. Simulated return loss against frequency for the microstrip-fed fork-shaped antenna with various inner radius R_2 ; other parameters are the same as listed in Table 1.

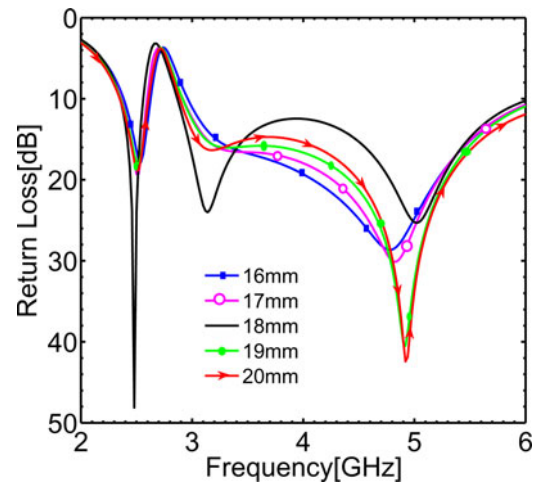


Fig. 8. Simulated return loss against frequency for the microstrip-fed fork-shaped antenna with various length W_g ; other parameters are the same as listed in Table 1.

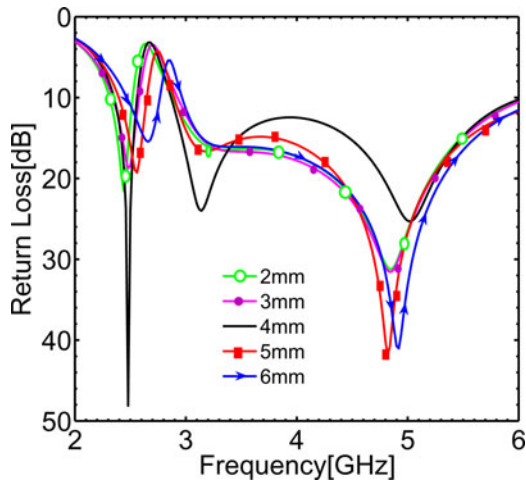


Fig. 6. Simulated return loss against frequency for the microstrip-fed fork-shaped antenna with various defect radius R_3 ; other parameters are the same as listed in Table 1.

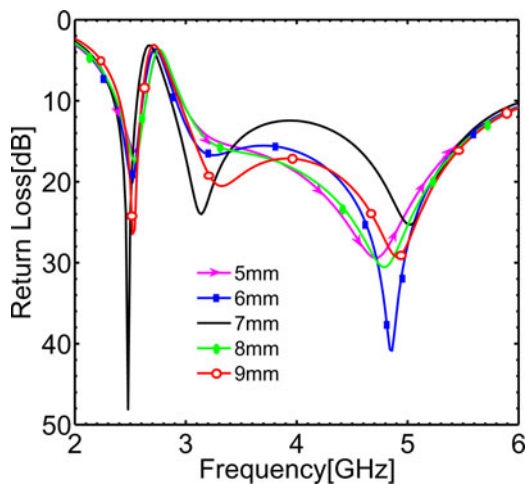


Fig. 7. Simulated return loss against frequency for the microstrip-fed fork-shaped antenna with various length L_{g1} ; other parameters are the same as listed in Table 1.

frequency of the second band. Thus, the optimum value for the width W_g of the ground is chosen as 18 mm. In Fig. 9, the effect of ground length L_g is shown. It is depicted from the graphs that L_g affects the impedance matching in both the bands. Thus, the optimum value of L_g is chosen as 2 mm.

IV. EXPERIMENTAL RESULTS AND DISCUSSION

Agilent N5230A vector network analyzer is used to measure the various parameters of the antenna such as impedance bandwidth, return loss, and radiation patterns. The measured and simulated return loss curves of the proposed antenna are shown in Fig. 10. It is observed from Fig. 10 that the simulated and measured results show relatively good agreement. The differences between the measured and simulated results are due to the effect of SMA (SubMiniature version A) connector soldering, substrate losses, and measurement circumstances. The differences may also be due to indeterminate error of

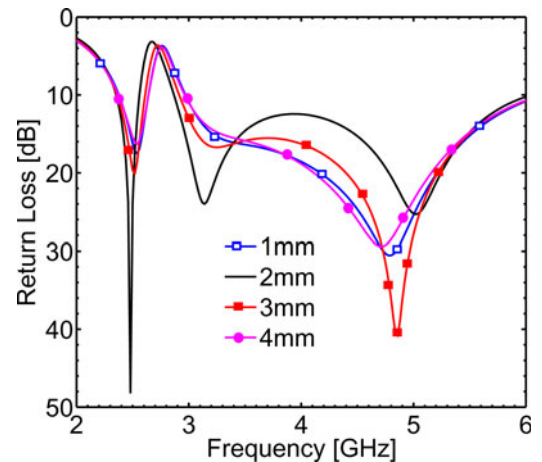


Fig. 9. Simulated return loss against frequency for the microstrip-fed fork-shaped antenna with various length L_g ; other parameters are the same as listed in Table 1.

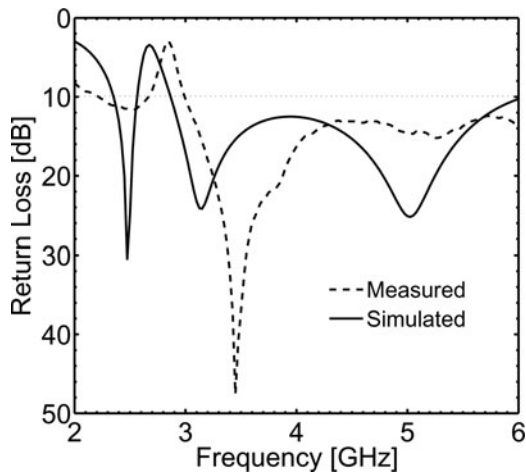


Fig. 10. Simulated and measured return loss of the proposed fork-shaped antenna.

the related dielectric constant and loss tangent during simulation. The antenna shows the measured bandwidth of 410 MHz (2.26–2.67) and 3.78 GHz (3.0–6.78 GHz) at lower and upper bands, respectively. It clearly covers both WLAN 2.4 GHz (2.4–2.484 GHz), 5.2/5.8 GHz (5.15–5.35 GHz/5.725–5.825 GHz), and WiMAX 2.5/3.5/5.5 GHz (2500–2690/3400–3690/5250–5850 MHz) bands. It is also observed from Fig. 11 that the radiation efficiency of the antenna is varied from 45 to 90%.

Figures 12(a)–12(c) show the two-dimensional far-field radiation patterns in the *H*- and *E*-planes at various sampling frequencies. It is observed that the antenna has nearly good omnidirectional radiation patterns at all operational frequencies in the *E*-plane (*xy*-plane). The radiation pattern of a monopole antenna is strongly affected by the finite-sized ground plane. For this conventional monopole antenna, the radiation pattern gets distorted in the *H*-plane due to radiation of a partial defected ground plane and mismatch of impedance. These patterns are suitable for application in most of the wireless communication equipment operating in WiMAX and WLAN bands, as expected.

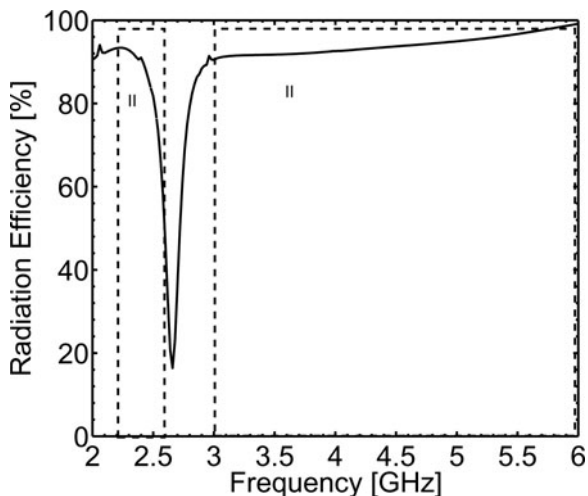


Fig. 11. Efficiency of the proposed fork-shaped antenna.

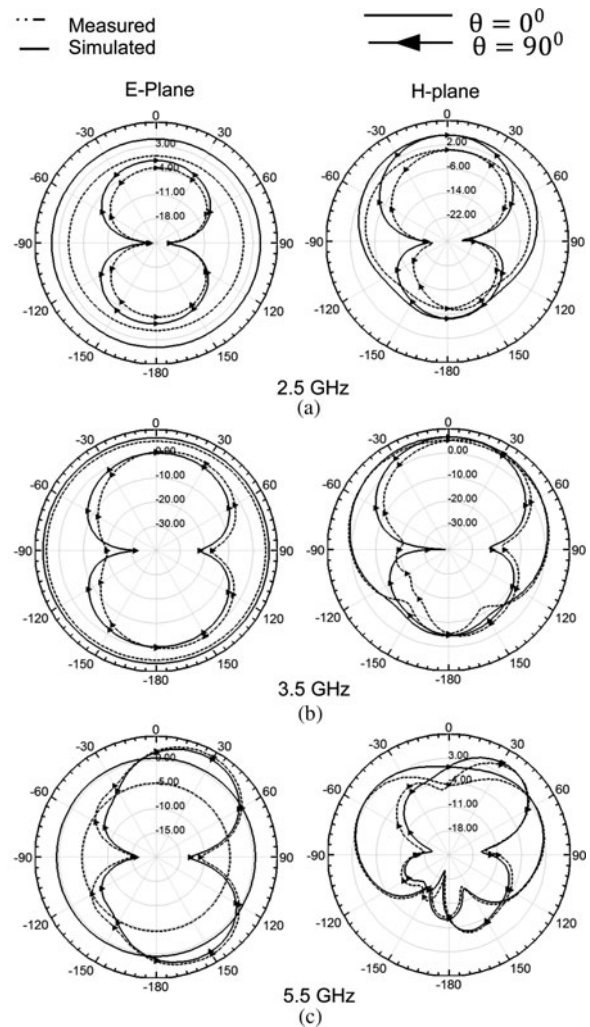


Fig. 12. Measured and simulated radiation patterns of the proposed fork-shaped antenna. *E*-field (*xy*-plane) and *H*-field (*yz*-plane).

V. CONCLUSION

A novel dual-band fork-shaped microstrip antenna is successfully proposed for Bluetooth, WLAN, and WiMAX applications. Measured results show a good agreement with the simulated results. The prototype with compact overall size of 24 mm × 35 mm × 1.6 mm achieves measured bandwidth of 410 MHz (2.26–2.67) and 3.78 GHz (3.0–6.78 GHz) at lower and upper bands, respectively. The proposed antenna is suitable for WLAN 2.4 GHz (2.4–2.484 GHz), 5.2/5.8 GHz (5.15–5.35 GHz/5.725–5.825 GHz), and WiMAX 2.5/3.5/5.5 GHz (2500–2690/3400–3690/5250–5850 MHz) bands applications.

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