

Design of a dual-frequency high-power microwave generator

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Abstract

A new direction for high-power microwave (HPM) development is to investigate devices capable of producing HPMs with a complicated spectrum. In recent years, some HPM sources with two stable and separate frequencies have been investigated theoretically and experimentally. However, many shortcomings still exist in these devices. Especially, the beam-wave interaction efficiency and the output microwave power are low in such devices. This paper proposes a novel dual-frequency HPM generator based on transition radiation. In the device, the electromagnetic fields are localized near the resonator cavities in the form of standing waves, and thus the interference between the different HPM components with different frequencies is weak. Compared with the existing dual-frequency devices, the new structure allows high beam-wave interaction efficiency and high output microwave power. As indicated in particle-in-cell simulation, with an electron beam of 500 kV voltage and 15.0 kA current guided by a magnetic field of 0.8 Tesla, an average power of 1.60 GW with a total power conversion efficiency of 21.3% is obtained, and the frequencies are 1.53 GHz and 3.29 GHz, respectively. Power level between two HPMs is comparable. The simulation results verify the feasibility of the dual-frequency HPM generator.

Keywords: Dual-frequency; High-power microwaves; Simulation; Transition radiation

INTRODUCTION

With the fast development of pulsed power technology (Liu *et al.*, 2006, 2007a, 2007b, 2008, 2009; Mesyats *et al.*, 2003; Yatsui *et al.*, 2005; Zou *et al.*, 2006) and intense electron beam generation from conventional accelerators (Li *et al.*, 2009a, 2009b, 2009c), high-power microwave (HPM) techniques have achieved a significant progress and microwave radiation with GW level have been obtained in experiments (Eltchaninov *et al.*, 2003; Korovin *et al.*, 2003; Xiao *et al.*, 2010). In pursuit of higher peak power and pulse energy, many new directions are becoming more and more attractive (Cao *et al.*, 2009; Ge *et al.*, 2010; Li *et al.*, 2010; Zhang *et al.*, 2010; Zhu *et al.*, 2010). One of these directions is to develop devices capable of producing HPMs with a complicated spectrum (Ginzburg *et al.*, 2002). In combination with this, some HPM sources with two stable and separate frequencies have been investigated theoretically and experimentally in recent years (Chen *et al.*, 2009; Fan *et al.*, 2007; Ju *et al.*, 2009; Wang *et al.*, 2010, 2011; Zhang *et al.*, 2008). Similar with the conventional circuit taxonomy, the existing

dual-frequency sources could be divided into two types, named cascading and parallel structures, respectively.

In the cascading structure, a single beam is used to generate the dual-frequency HPMs (Chen *et al.*, 2009; Wang *et al.*, 2010). Such devices are generally of low beam-wave interaction efficiency due to the interference between the two HPM components with different frequencies. Simultaneously, the transportation of the microwave power generated in one section is greatly influenced by the other section, and thus the devices are probably of low power extraction efficiency. Therefore, there are many difficulties for the cascading structure to design the proper dynamic structure.

In the parallel structure, the beam-wave interaction regions for generation of dual-frequency HPMs are separated and two beams emitted by the same cathode are used to generate their respective microwave (Fan *et al.*, 2007; Ju *et al.*, 2009; Wang *et al.*, 2010, 2011; Zhang *et al.*, 2008). However, such devices greatly depend on the design of the cathode distributing the beam currents. Besides, the microwaves with different frequencies are probably radiated at different time due to inconsistent excitation and saturation time. Because of independence of the HPMs generation, the design of the parallel structure is in essence consistent with the HPM source with single frequency.

Considering these reasons, it would be valuable to seek a HPM generator both with high output microwave power and simultaneous microwave radiation at dual frequencies. In this

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paper, a dual-frequency HPM generator based on transition radiation is put forward. The electromagnetic fields in the device are localized near the resonator cavities in the form of standing waves and thus the interference between the different HPM components with different frequencies is greatly degraded. Simultaneously, the time synchronism can be realized because the radiated microwaves are from a single electron beam with dual-frequency modulation. Owing to the coaxial structure, the new device also allows high input electric power and high output microwave power.

DESCRIPTION OF THE PRINCIPLE

Radiation Mechanism and Physical Model

The main varieties of microwave tubes are classified into three groups, according to the fundamental radiation mechanism involved: Cherenkov, transition, or bremsstrahlung radiation (Gold & Nusunovich, 1997). As is known, in microwave tubes based on transition radiation, the microwave fields are localized within the short-gap cavities (bunchers) and coherent radiation is produced by decelerating the electron bunches in the output cavity. Therefore, the fields in one buncher have a weak influence on the fields in the other buncher. An electron beam in turn modulated by two types of bunchers with different frequencies probably stimulates the dual-frequency HPMs in the vicinity of output cavity.

Figure 1a gives a three-dimensional general schematic of the designed dual-frequency HPM generator. Figure 1b is the corresponding simulation model, which is symmetric about the z -axis. As shown in the figures, the device mainly consists of five parts: an annular cathode, two sequent dual-cavity bunchers, respectively, corresponding to two different frequencies, a dual-cavity extractor, and a coaxial output structure. The resonance frequencies of the TM_{01} modes in the two bunchers mainly depend on the radial dimensions and can be given approximately by

$$f_0(\text{GHz}) = \frac{c}{2\Delta r} = \frac{15}{\Delta r(\text{cm})}(\text{GHz}), \quad (1)$$

and thus they can be chosen by adjusting the radii of the bunchers. The dual-cavity extractor includes two cavities to stimulate the respective microwaves. The first cavity has the same radial dimension with the buncher c (Fig. 1) and the second cavity has the same radial dimension with the buncher b (Fig. 1), respectively.

The basic principle of the proposed model is as follows. The annular electron beam emitted by the cathode transmits axially due to the restriction of the guiding magnetic field, and the velocity of electrons is modulated in succession by the electromagnetic fields with different frequencies when they pass through the two sequent bunchers. Finally, HPMs with two separate frequencies are simultaneously stimulated in the region of the dual-cavity extractor and extracted by the coaxial output structure. Compared with the

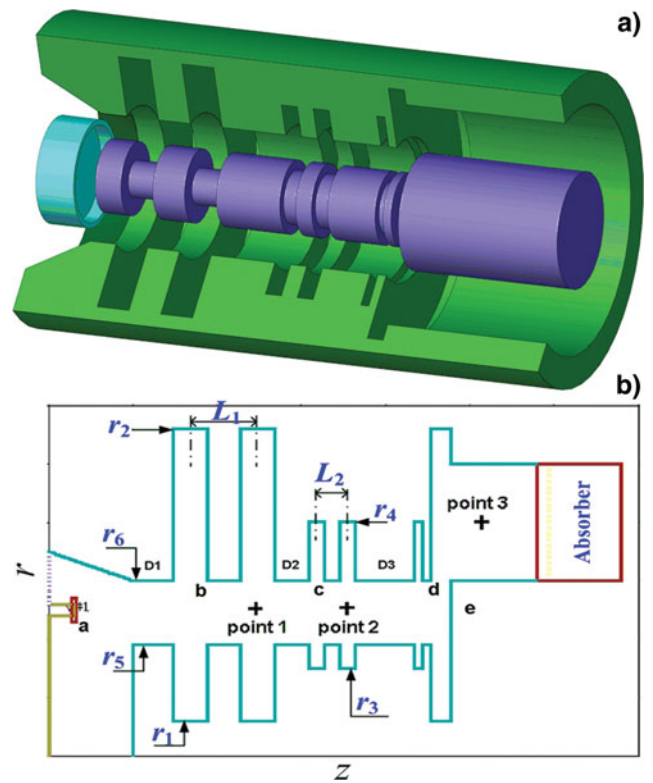


Fig. 1. (Color online) The three-dimensional schematic (a) and simulation model (b) of the designed dual-frequency high-power microwave generator. In figure (b), the characters a, b, c, d, and e represent the annular cathode, the first and second dual-cavity bunchers, the dual-cavity extractor, and the beam dump, respectively.

existing dual-frequency HPM generators, the designed device has the following merits: (1) The interference between the two HPM components with different frequencies is very weak because the fields are localized within the corresponding cavities. (2) The radiated microwaves are stimulated by the single modulated bunches in the vicinity of output cavity, and thus the time synchronism could be satisfied.

Correlation between Two Frequencies

Although the interference between the two bunchers with different frequencies is very weak, the previous bunches modulated by the first buncher is probably destroyed by the second buncher with the other operating frequency. Such a case could be improved by choosing two proper operating frequencies. Further analysis shows that the operating frequency in the second buncher should be multiple of the operating frequency in the first buncher.

Figure 2 denotes bunching of an electron beam at frequencies f and $2f$. Obviously, the centers of bunches corresponding to frequency f is simultaneously located at the phase corresponding to zero or gain or absorption when the beam passes through the field with frequency $2f$. Therefore, the latter would not destroy the previous bunches. Besides, the

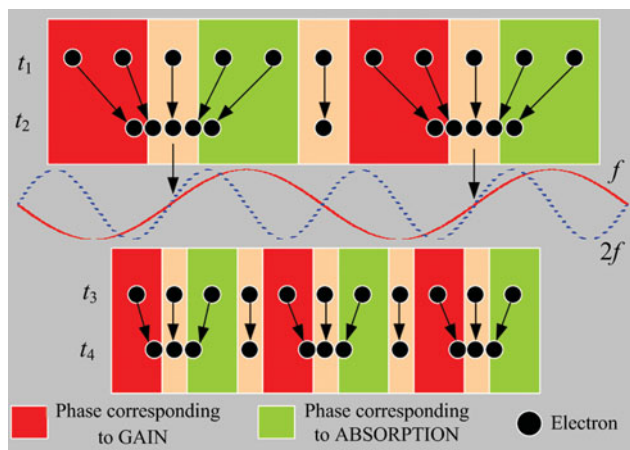


Fig. 2. (Color online) Process of beam bunching at frequency f and $2f$.

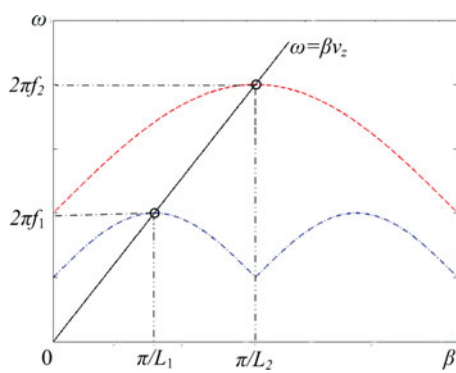


Fig. 3. (Color online) Dispersion relations corresponding to two different HPM components. Frequencies supported by the devices based on transition radiation are circled.

direct-current component of the beam can be modulated by the field at frequency $2f$. Eventually, the dual frequency modulations of the electron beam are realized.

Figure 3 describes the dispersion relations (Lemke, 1992), respectively, corresponding to two HPM components with different frequencies. In the figure, ω and β , respectively, denote the angular frequency and the mode wave number, and L_1 and L_2 , respectively, represent the structure periods of the buncher b and buncher c (Fig. 1). The intersections of the beam line and the modes represent the working points of the dual-frequency HPM generator. From the figure, it is easy to yield

$$v_z = \frac{\omega}{\beta} = 2f_1 L_1 = 2f_2 L_2. \quad (2)$$

In case that the frequency f_2 is twice the frequency f_1 , the structure period L_2 should be half of the structure period L_1 .

SIMULATION RESULTS

Cold Test

Based on the above analyses, the main structure parameters are chosen as follows: $r_1 = 1.2$ cm, $r_2 = 11.2$ cm, $r_3 =$

3.0 cm, $r_4 = 8.0$ cm, $r_5 = 3.8$ cm, $r_6 = 6.0$ cm, $L_1 = 8.0$ cm, and $L_2 = 4.0$ cm (Fig. 1). According to Eq. (1), the operating frequencies for TM_{01} mode of the two bunchers are about 1.5 GHz (L-band) and 3.0 GHz (S-band), respectively. For the TM_{0n} mode in the coaxial structure, the axial electric field E_{zn} can be expressed as

$$E_{zn}(r, z) = [A_n J_n(k_c r) + B_n Y_n(k_c r)] e^{-j\beta z} \quad (n = 1, 2, 3, \dots), \quad (3)$$

where J_n and Y_n , respectively, represent the first and second kind Bessel functions with order n . In the coaxial drift-tube of inner radius r_5 and outer radius r_6 , the field E_{zn} satisfies the boundary conditions: $E_{zn}(r_5, z) = 0$ and $E_{zn}(r_6, z) = 0$. That is,

$$\begin{cases} A_n J_n(k_c r_5) + B_n Y_n(k_c r_5) = 0 \\ A_n J_n(k_c r_6) + B_n Y_n(k_c r_6) = 0 \end{cases} \quad (4)$$

Combining the above two expressions gives

$$\frac{J_n(k_c r_5)}{Y_n(k_c r_5)} = \frac{J_n(k_c r_6)}{Y_n(k_c r_6)} \quad (5)$$

Setting $n = 1$ and substituting the given values into the above expression, the cut-off frequency of TM_{01} mode is about 6.87 GHz in the drift-tube. Therefore, the TM_{01} modes at L-band and S-band can be effectively cut off in the drift-tube.

Figure 4 shows the distributions of the electric fields corresponding to the two bunchers. The Eigen frequencies are 1.5 GHz and 3.2 GHz, respectively. Obviously, the two bunchers share a common radial area, where intense axial electric fields exist simultaneously and an electron beam is allowed to pass through. Figure 5 gives the radial profiles of the axial electric fields at the second cavity center of the L-band and S-band bunchers (corresponding to point 1 and point 2 in Fig. 1). As indicated in the figure, the axial fields can maintain high level in the channel of beam transportation and theoretically this is very advantageous to

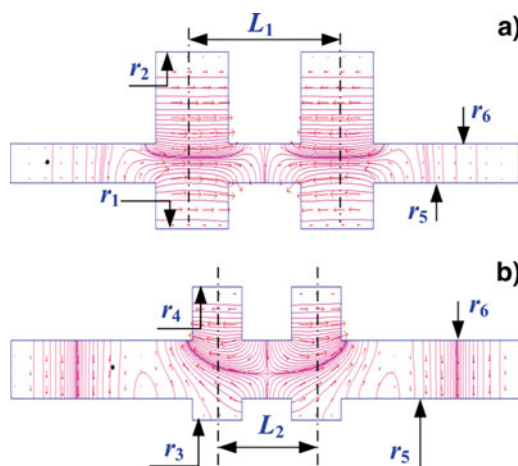


Fig. 4. (Color online) Electric field distributions in (a) L-band buncher and (b) S-band buncher.

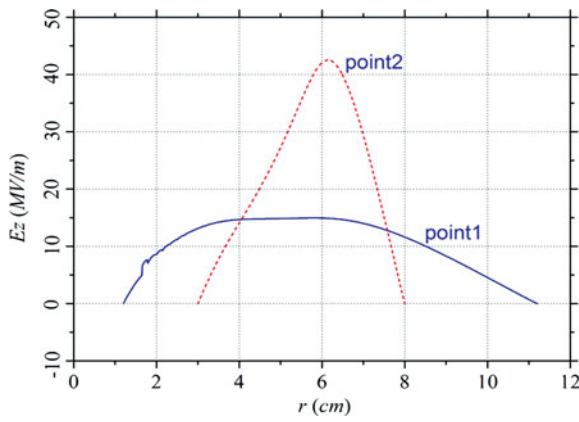


Fig. 5. (Color online) Radial profiles of eigen-fields in the L-band and S-band bunchers.

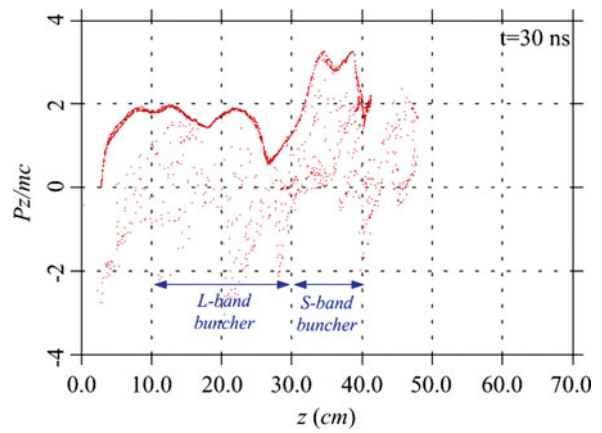


Fig. 7. (Color online) Phase-space plot for the dual-frequency HPM generator.

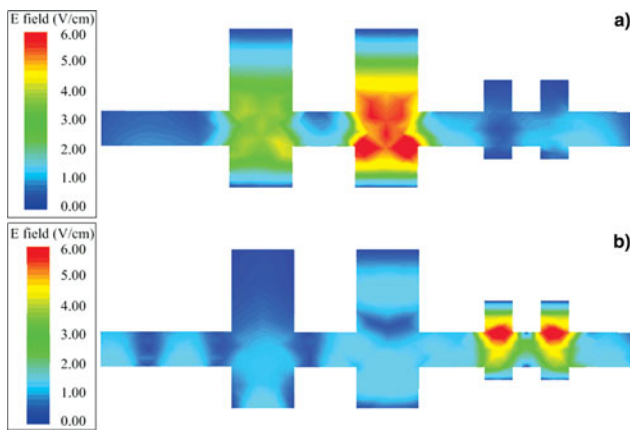


Fig. 6. (Color online) Electric field distributions in the cascading L-band and S-band bunchers. (a) and (b) correspond to the L-band and S-band operating modes, respectively.

beam modulation. As a first approximation, the average radius of the electron beam would be that for which the axial electric fields are maximum. The practical location that electron beam passes through should be optimized in combination with the two bunchers.

Figure 6 denotes that the L-band and S-band electromagnetic fields can be well localized in their respective buncher, which further verifies the independence of the two bunchers. Consequently, the interference between the different HPM components with different frequencies is greatly degraded.

Hot Test

Last, the dual-frequency HPM generator is optimized with the particle-in-cell simulation code KARAT (Tarakanov, 1998). With an electron beam of 500 kV voltage and 15.0 kA current guided by a magnetic field of 0.8 Tesla, the main simulation results are given as follows.

The phase-space plot for the dual-frequency HPM generator is displayed in Figure 7, which illustrates the energy spread of the modulated beam electrons. It can be seen

from this figure that the electron beam emitted by the cathode is accelerated in the diode region by the direct-current voltage, then in turn modulated by the electromagnetic fields in the L-band and S-band bunchers. Such an electron beam would stimulate the HPMs corresponding to the two bands in the vicinity of the dual-cavity extractor and be eventually collected by the beam dump.

Figure 8 shows the variation of the electric fields versus time and corresponding frequency spectrum at point 1 (Fig. 1). This exhibits electromagnetic fields stimulated by

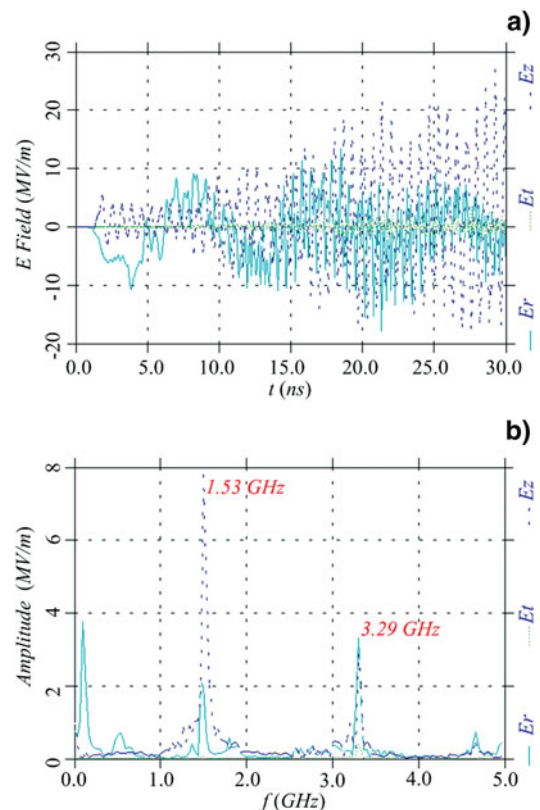


Fig. 8. (Color online) Variation of electric fields versus time and corresponding frequency spectrum at point 1 (Fig. 1).

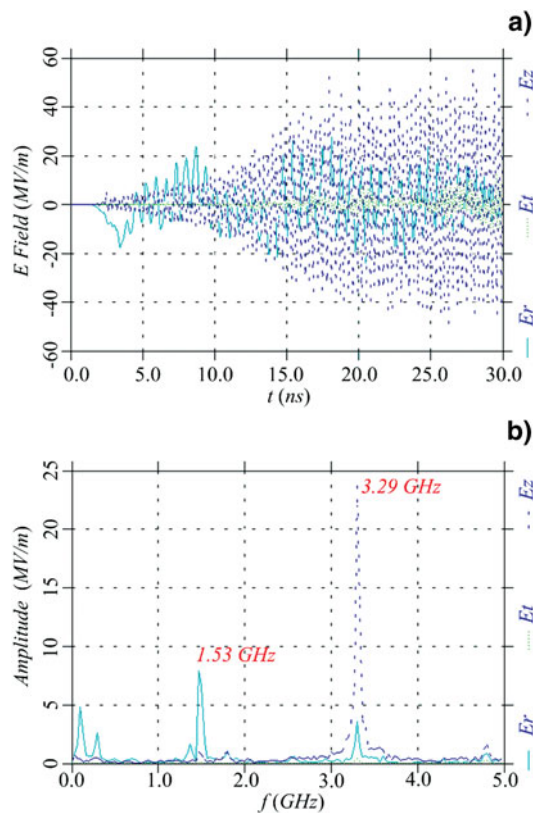


Fig. 9. (Color online) Variation of electric fields *versus* time and corresponding frequency spectrum at point 2 (Fig. 1).

the electron beam in the L-band buncher. The main frequency of the axial electric field modulating the beam is about 1.53 GHz, which is slightly greater than the Eigen frequency 1.5 GHz because of a loaded electron beam. Besides, the frequency of the second harmonic is close to the Eigen frequency of the S-band buncher and such a microwave component is probably helpful to stimulate the S-band microwave. As indicated in the figures, some unexpected modes also exist in the form of the transverse electromagnetic modes (TEM), which seem to be inevitable due to the coaxial inner conductor. However, the TEM modes cannot interact

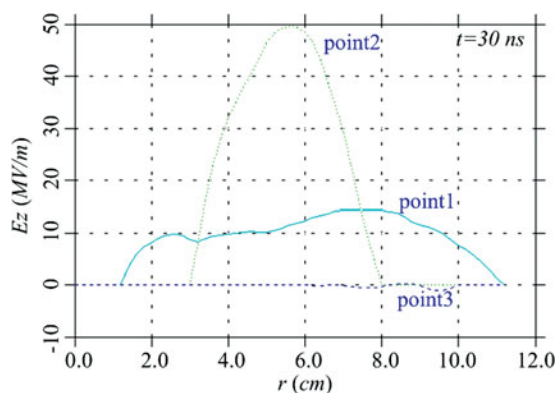


Fig. 10. (Color online) Radial profiles of modulating fields in the L-band and S-band bunchers.

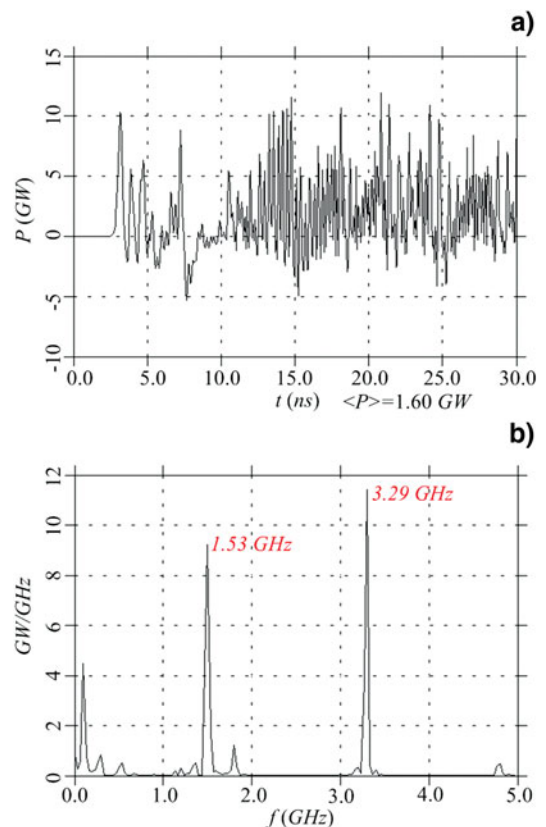


Fig. 11. (Color online) Variation of the microwave power *versus* time and corresponding frequency spectrum at point 3 (Fig. 1).

with the electron beam with axial velocity and thus they hardly influence the beam modulation.

Figure 9 shows the variation of the electric fields *versus* time and corresponding frequency spectrum at point 2 (Fig. 1). Obviously, the electromagnetic fields are stimulated in the S-band buncher and the main frequency of the axial electric field is 3.29 GHz, which is approximately twice the frequency of the L-band microwave. As mentioned above, such a component would not destroy bunches modulated by the first buncher. Also, some unexpected TEM modes exist in the S-band buncher and they hardly have influence over the beam modulation.

As shown in Figure 10, the radial structures of the axial electric fields in the L-band and S-band bunchers (point 1 and point 2 in Fig. 1) are basically in accordance with the results shown in Figure 5. That is to say, the operating modes are indeed the TM_{01} modes. Due to the loaded electron beam, the maximal values are partly shifted compared with the previous results (Fig. 5). However, in the actual beam path ($r \approx 5.0$ cm), the modulating fields at both the L-band and the S-band still maintain high level. No obvious axial components are observed at point 3 (Fig. 1) and the electromagnetic fields exist in the coaxial output structure mainly in the form of TEM waves.

Figure 11 shows the variation of the microwave power *versus* time and corresponding frequency spectrum at point

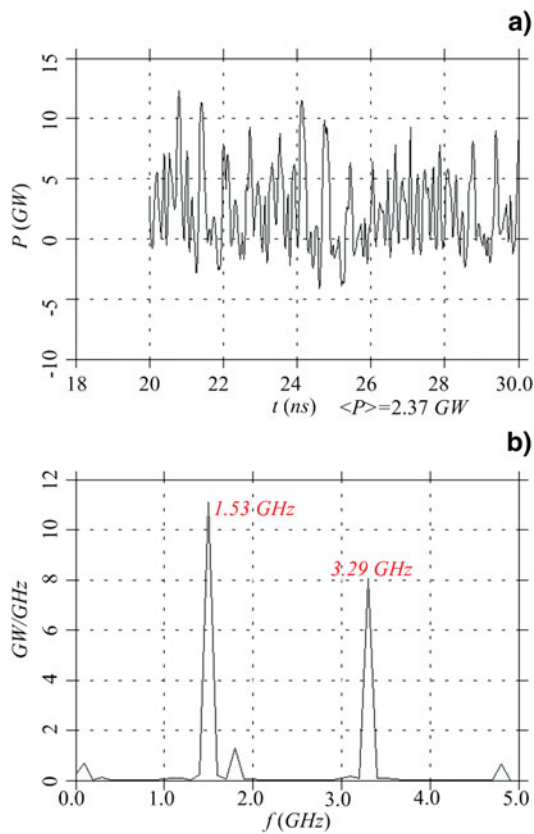


Fig. 12. (Color online) Output power variation in the range from 20 ns to 30 ns and its corresponding frequency spectrum.

3 (Fig. 1). As shown in Figure 11a, a total average power of 1.60 GW is obtained in the coaxial output structure and the efficiency is about 21.3%. The corresponding spectrum is given in Figure 11b and the dominant frequencies are 1.53 GHz and 3.29 GHz, respectively. The spectrum indicates that the power level of the two bands is comparative. Besides, from the spectrum, we can also see that there exist some unexpected frequency components, especially the frequency about 100 MHz, which takes away a lot of energy. Such frequency components are probably stimulated in the process of beam voltage rise, corresponding to the time domain from about 3 ns to 10 ns in Figure 11a. Figure 12 gives the output power variation in the range from 20 ns to 30 ns and its corresponding frequency spectrum. Obviously, after the microwaves get to saturation, the unexpected

frequency components are significantly degraded, and the total power of the output microwave is about 2.37 GW with dominant frequencies 1.53 GHz and 3.29 GHz.

COMPARISON WITH OTHER DUAL-FREQUENCY DEVICES

Table 1 lists the recent simulation results of some existing dual-frequency HPM sources (Chen *et al.*, 2009; Fan *et al.*, 2007; Ju *et al.*, 2009; Wang *et al.*, 2010), including the proposed device in this paper. Due to higher input electric power, one can see clearly that the devices based on dual beams have higher output power level compared with the devices based on a single beam. However, as mentioned above, the two microwaves generated by the devices based on dual beams are probably radiated at different time because of inconsistent excitation and saturation time.

The really attractive dual-frequency HPM sources should be those devices allowing two microwaves simultaneous radiation. Obviously, the dual-frequency device based a single beam is more feasible to realize the time synchronism. However, as shown in Table 1, the power level of such devices is still dissatisfactory up to now. The proposed device in this paper possesses the highest power conversion efficiency compared with the other devices shown in the table. With almost the same input electric power, the output power of the proposed device is higher than that of the dual-frequency backward wave oscillator (BWO) based on dual beams (marked by superscript asterisk in Table 1). Even if the input power is only half of that of the magnetically insulated line oscillator (MILO) based on a single beam (marked by superscript plus sign in Table 1), the output power of the proposed device is still higher.

CONCLUSIONS

In this paper, a dual-frequency high-power microwave generator based on transition radiation is designed and investigated theoretically. In the device, the electromagnetic fields at different frequencies are localized in their respective buncher and thus the interference between the two wave bands is greatly degraded. Besides, the electromagnetic fields at different frequencies are radiated into the outer space by the same electron beam at the same time.

Table 1. Simulation results of some existing dual-frequency HPM devices

Source	Beam type	Input power (GW)	Output microwave (GW)	Efficiency	Frequency (GHz)	Magnetic field (T)
MILO	Dual	50.02	5.9	11.80%	7.6/9.26	—
MILO-VCO	Dual	32	5.22	16.30%	1.33/3.91	—
BWO*	Dual	7.67	1.4	18.30%	4.6/8.4	2
BWO	Single	3.25	0.38	11.70%	5.48/9.6	2
MILO ⁺	Single	15.96	1.3	8.10%	1.27/1.49	—
Proposed Device	Single	7.5	1.6	21.3%	1.53/3.29	0.8

Further, the dual-frequency HPM generator is optimized with the particle-in-cell simulation code KARAT. With an electron beam of 500 keV and 15.0 kA guided by a magnetic field of 0.8 Tesla, an average power of 1.60 GW with a total power conversion efficiency of 21.3% is obtained and the frequencies are 1.53 GHz and 3.29 GHz, respectively. As indicated by the simulation results, the design of the dual-frequency HPM generator is feasible.

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